

UNITED STATES PATENT APPLICATION

FOR

A PLANAR LIGHTWAVE CIRCUIT HAVING AN INTEGRATED
DISPERSION COMPENSATOR USING A FOURIER FILTER

Inventors:

Koorosh Aflatooni
Tony Ticknor

Prepared by:
WAGNER, MURABITO & HAO, LLP
Two North Market Street
Third Floor
San Jose, California 95113

A PLANAR LIGHTWAVE CIRCUIT HAVING AN INTEGRATED
DISPERSION COMPENSATOR USING A FOURIER FILTER

5

FIELD OF THE INVENTION

The present invention relates generally to the signal optimization of modern fiber-optic communications networks. More particularly, the present invention relates to a method and system for correcting chromatic dispersion
10 using planar lightwave circuits.

BACKGROUND OF THE INVENTION

Planar lightwave circuits comprise fundamental building blocks for the newly emerging, modern fiber-optic communications infrastructure. Planar
15 lightwave circuits are innovative devices configured to transmit light in a manner analogous to the transmission of electrical currents in printed circuit boards and integrated circuit devices. Examples include arrayed waveguide grating devices, integrated wavelength multiplexers/demultiplexers, optical switches, optical modulators, wavelength-independent optical couplers, and
20 the like.

Planar lightwave circuits (PLCs) generally involve the provisioning of a series of embedded optical waveguides upon a semiconductor substrate (e.g., silicon), with the optical waveguides fabricated from one or more silica

glass layers, formed on an underlying semiconductor substrate. Fabrication techniques required for manufacturing PLCs using silica glass is a newly emerging field. Electronic integrated circuit type (e.g., CMOS) semiconductor manufacturing techniques have been extensively developed to aggressively

5 address the increasing need for integration in, for example, the computer industry. This technology base is currently being used to make PLCs. By using manufacturing techniques closely related to those employed for silicon integrated circuits, a variety of optical circuit elements can be placed and interconnected on the surface of a silicon wafer or similar substrate. This

10 technology has only recently emerged and is advancing rapidly with leverage from more mature tools of the semiconductor-processing industry.

PLCs are constructed with a number of waveguides precisely fabricated and laid out across a silicon wafer. A conventional optical waveguide comprises an un-doped silica bottom clad layer, with at least one waveguide core formed thereon, and a cladding layer covering the waveguide core, wherein a certain amount of at least one dopant is added to both the waveguide core and the cladding layer so that the refractive index of the waveguide core is higher than that of the cladding layer. Fabrication of

15 conventional optical waveguides involves the formation of a silica layer as the bottom clad, usually grown by thermal oxidation upon a silicon semiconductor wafer. The core layer is a doped silica layer, which is deposited by either plasma-enhanced chemical vapor deposition (PECVD) or flame hydrolysis deposition (FHD). An annealing procedure then is applied

20 to this core layer (heated above 1000C). The waveguide pattern is subsequently defined by photolithography on the core layer, and reactive ion etching (RIE) is used to remove the excess doped silica to form one or more

25

waveguide cores. A top cladding layer is then formed through a subsequent deposition process. Finally, the wafer is cut into multiple PLC dies and the dies are packaged according to their particular applications.

5 Prior art Figure 1 shows a cross-section view of a conventional planar optical waveguide. As depicted in Figure 1, the planar optical waveguide includes a doped SiO₂ glass core 10 formed over a SiO₂ bottom cladding layer 12 which is on a silicon substrate 13. A SiO₂ top cladding layer 11 covers both the core 10 and the bottom cladding layer 12. As described above, the refractive index of the core 10 is higher than that of the cladding layers 11 and 12. Consequently, optical signals are confined axially within core 10 and propagate lengthwise through core 10. The SiO₂ glass core 10 is typically doped with Ge or P to increase its refractive index. In many types of PLC devices, a large number of cores (e.g., 40 or more) are used to implement complex fiber-optic functions, such as, for example, arrayed waveguide grating multichannel multiplexers and de-multiplexers.

10

15

Thus, PLCs comprise fundamental building blocks for the modern fiber-optic communications infrastructure. The PLCs provide the means for organizing and concentrating optical signals for transmission at one end of an optical fiber and the means for extracting and detecting optical signals received at the other end of the optical fiber.

20

There exists a problem however, with signal dispersion, or more specifically, chromatic dispersion, in the transmission of optical signals across long distances through fiber-optic cables (e.g., bundles of optical fibers). As is well known, chromatic dispersion is an important issue in high-

25

speed fiber optic communication networks. As the bit rate of a transmission system increases, the susceptibility to chromatic dispersion increases. For example, in a fiber-optic transmission system functioning at 40Gb per second per channel, the system can only tolerate a dispersion in order of 10 ps/nm.

5

Prior art Figure 2 shows a diagram of a transmitted optical pulse on the left and the corresponding received optical pulse on the right, and prior art Figure 3 shows a diagram of a fiber-optic transmitter 21 transmitting the optical pulse across a distance of fiber-optic cable to a fiber-optic receiver 22.

- 10 As depicted in Figure 1, the transmitted optical pulse undergoes chromatic dispersion, wherein the power of the optical pulse (on the vertical axis) is spread with respect to time (on the horizontal axis). This occurs as the transmitted optical pulse propagates through long distances of optical fiber, as shown in Figure 2. The different frequency components of the transmitted pulse propagate through the optical fiber at different speeds. Thus, the relatively square profile of the transmitted pulse on the left becomes the dispersed profile of the received pulse on the right after propagation through some distance of optical fiber (e.g., 100 kilometers or more).
- 15

- 20 Prior art Figure 4 shows consecutive pulses of a signal received at the receiver 22 with respect to the same pulses transmitted from the transmitter 21. The problem occurs in the receiver 22 when the receiver 22 tries to sample an incoming pulse train of a signal. As depicted in Figure 3, dispersion causes the power of consecutive pulses to blend in with one another. This can cause difficulty when the receiver 22 tries to sample the pulses to determine their logic level. Instead of being distinct square profile pulses as transmitted from the transmitter 21, the pulses are dispersed and
- 25

blended when received by the receiver 22, making them difficult to reliably sample.

As described above, as the bit rate of a transmission system increases,
5 the susceptibility to chromatic dispersion increases. High transmission frequencies means less time between the pulses. For example, in a fiber-optic transmission system functioning at 40Gb per second, the system can only tolerate a dispersion in the order of 10 ps/nm before the dispersion makes the reliable detection and sampling of individual pulses virtually impossible.
10 Additionally, the dispersion characteristics of the modern fiber-optic communications network change as the network configuration changes (e.g., as new network nodes are added, new cables added, new channels are multiplexed, and the like). Thus, any attempt at compensating for dispersion must be able to account for such changing network conditions.

15 Prior art solutions for dynamically compensating for chromatic dispersion have involved the use of ring resonators (e.g., IIR type filters) and variable phase shifters. These solutions have a critical drawback in that the fabrication of ring resonators and variable phase shifters have a number of
20 critical dimensions. The critical dimensions make such devices very difficult to reliably manufacture.

Thus what is required is a solution capable of balancing the dispersion introduced through various building blocks of a communications network.

25 Additionally, since the characteristics of the communications network change as the network reconfigures, the required solution should have a dispersion compensation means capable of handling not only static network conditions,

but also dynamic, changing network conditions. The required solution should be more easily manufactured than prior art solutions and should not require very tight critical dimension control. The present invention provides a novel solution to the above requirements.

5

95
96
97
98
99
100
101
102
103
104
105
106
107
108
109
110
111
112
113
114
115
116
117
118
119
120
121
122
123
124
125
126
127
128
129
130
131
132
133
134
135
136
137
138
139
140
141
142
143
144
145
146
147
148
149
150
151
152
153
154
155
156
157
158
159
160
161
162
163
164
165
166
167
168
169
170
171
172
173
174
175
176
177
178
179
180
181
182
183
184
185
186
187
188
189
190
191
192
193
194
195
196
197
198
199
200
201
202
203
204
205
206
207
208
209
210
211
212
213
214
215
216
217
218
219
220
221
222
223
224
225
226
227
228
229
230
231
232
233
234
235
236
237
238
239
240
241
242
243
244
245
246
247
248
249
250
251
252
253
254
255
256
257
258
259
260
261
262
263
264
265
266
267
268
269
270
271
272
273
274
275
276
277
278
279
280
281
282
283
284
285
286
287
288
289
290
291
292
293
294
295
296
297
298
299
300
301
302
303
304
305
306
307
308
309
310
311
312
313
314
315
316
317
318
319
320
321
322
323
324
325
326
327
328
329
330
331
332
333
334
335
336
337
338
339
340
341
342
343
344
345
346
347
348
349
350
351
352
353
354
355
356
357
358
359
360
361
362
363
364
365
366
367
368
369
370
371
372
373
374
375
376
377
378
379
380
381
382
383
384
385
386
387
388
389
390
391
392
393
394
395
396
397
398
399
400
401
402
403
404
405
406
407
408
409
410
411
412
413
414
415
416
417
418
419
420
421
422
423
424
425
426
427
428
429
430
431
432
433
434
435
436
437
438
439
440
441
442
443
444
445
446
447
448
449
450
451
452
453
454
455
456
457
458
459
460
461
462
463
464
465
466
467
468
469
470
471
472
473
474
475
476
477
478
479
480
481
482
483
484
485
486
487
488
489
490
491
492
493
494
495
496
497
498
499
500
501
502
503
504
505
506
507
508
509
510
511
512
513
514
515
516
517
518
519
520
521
522
523
524
525
526
527
528
529
530
531
532
533
534
535
536
537
538
539
540
541
542
543
544
545
546
547
548
549
550
551
552
553
554
555
556
557
558
559
559
560
561
562
563
564
565
566
567
568
569
569
570
571
572
573
574
575
576
577
578
579
579
580
581
582
583
584
585
586
587
588
589
589
590
591
592
593
594
595
596
597
598
599
599
600
601
602
603
604
605
606
607
608
609
609
610
611
612
613
614
615
616
617
618
619
619
620
621
622
623
624
625
626
627
628
629
629
630
631
632
633
634
635
636
637
638
639
639
640
641
642
643
644
645
646
647
648
649
649
650
651
652
653
654
655
656
657
658
659
659
660
661
662
663
664
665
666
667
668
669
669
670
671
672
673
674
675
676
677
678
679
679
680
681
682
683
684
685
686
687
688
689
689
690
691
692
693
694
695
696
697
698
699
699
700
701
702
703
704
705
706
707
708
709
709
710
711
712
713
714
715
716
717
718
719
719
720
721
722
723
724
725
726
727
728
729
729
730
731
732
733
734
735
736
737
738
739
739
740
741
742
743
744
745
746
747
748
749
749
750
751
752
753
754
755
756
757
758
759
759
760
761
762
763
764
765
766
767
768
769
769
770
771
772
773
774
775
776
777
778
779
779
780
781
782
783
784
785
786
787
788
789
789
790
791
792
793
794
795
796
797
798
799
799
800
801
802
803
804
805
806
807
808
809
809
810
811
812
813
814
815
816
817
818
819
819
820
821
822
823
824
825
826
827
828
829
829
830
831
832
833
834
835
836
837
838
839
839
840
841
842
843
844
845
846
847
848
849
849
850
851
852
853
854
855
856
857
858
859
859
860
861
862
863
864
865
866
867
868
869
869
870
871
872
873
874
875
876
877
878
879
879
880
881
882
883
884
885
886
887
888
889
889
890
891
892
893
894
895
896
897
898
899
899
900
901
902
903
904
905
906
907
908
909
909
910
911
912
913
914
915
916
917
918
919
919
920
921
922
923
924
925
926
927
928
929
929
930
931
932
933
934
935
936
937
938
939
939
940
941
942
943
944
945
946
947
948
949
949
950
951
952
953
954
955
956
957
958
959
959
960
961
962
963
964
965
966
967
968
969
969
970
971
972
973
974
975
976
977
978
979
979
980
981
982
983
984
985
986
987
988
989
989
990
991
992
993
994
995
996
997
998
999
999
1000

SUMMARY OF THE INVENTION

The present invention provides a solution capable of balancing the dispersion introduced through various building blocks of a communications network. The present invention provides dynamic dispersion compensation capable of handling not only static network conditions, but also changing network conditions. The present invention is more easily manufactured than prior art solutions and does not require very tight critical dimension control.

In one embodiment, the present invention is implemented as a planar lightwave circuit (PLC) having an integrated dispersion compensator fabricated therein. The PLC includes an input for receiving a fiber optic signal from, for example, a fiber-optic communications network. The input couples the signal to a Fourier filter built into the PLC. The filter is configured to add a phase compensation (e.g., a phase compensation profile) to the signal to correct a chromatic dispersion of the signal. An output is coupled to transmit the dispersion compensated signal from the Fourier filter to other components on the PLC, such as for example, an integrated arrayed waveguide grating, or to other external devices, such as, for example, an external optical receiver.

20

The Fourier filter can be implemented using a tap delay filter. The tap delay filter can be implemented by using a plurality of delay lines for implementing the phase compensation for the signal. The delay lines can be implemented using Mach Zehnder couplers, wherein the Mach Zehnder couplers are configured to distribute power from the incoming signal between the delay lines and to recombine the power from the delay lines to generate

- the dispersion compensated signal. A plurality of thermal optic phase shifters can be coupled to the delay lines to generate the phase compensation. The thermal optic phase shifters provide dynamic phase compensation to account for changes in network conditions. The Fourier filter can be
- 5 manufactured using standard PLC manufacturing techniques and does not require excessively tight critical dimension control.

Lightwave Technologies, Inc.
10000 North 100th Street
Milwaukee, WI 53223-3767
(414) 766-1000
(800) 343-4444

BRIEF DESCRIPTION OF THE DRAWINGS

The present invention is illustrated by way of example and not by way of limitation, in the Figures of the accompanying drawings and in which like reference numerals refer to similar elements and in which:

5

Prior art Figure 1 shows a cross-section view of a conventional planar optical waveguide fabricated using a silica glass substrate.

10 Prior art Figure 2 shows a diagram of a transmitted optical pulse on the left and the received optical pulse on the right.

Prior art Figure 3 shows a diagram of a fiber-optic transmitter transmitting an optical pulse train across a distance of fiber-optic cable to a fiber-optic receiver.

15

Prior art Figure 4 shows consecutive pulses of a signal received at the receiver with respect to the same pulses transmitted from the transmitter.

20 Figure 5 shows a fiber-optic communication system in accordance with one embodiment of the present invention.

Figure 6 shows a diagram of two signal pulses received at the filter and the two signal pulses after filtering by the filter in accordance with one embodiment of the present invention.

25

Figure 7 shows a tap delay filter in accordance with one embodiment of the present invention.

4700 4701 4702 4703 4704 4705 4706 4707 4708 4709 4710 4711 4712 4713 4714 4715 4716 4717 4718 4719 4720 4721 4722 4723 4724 4725 4726 4727 4728 4729 4730 4731 4732 4733 4734 4735 4736 4737 4738 4739 4740 4741 4742 4743 4744 4745 4746 4747 4748 4749 4750 4751 4752 4753 4754 4755 4756 4757 4758 4759 4760 4761 4762 4763 4764 4765 4766 4767 4768 4769 4770 4771 4772 4773 4774 4775 4776 4777 4778 4779 4770 4771 4772 4773 4774 4775 4776 4777 4778 4779

DETAILED DESCRIPTION OF THE INVENTION

Reference will now be made in detail to the embodiments of the invention, examples of which are illustrated in the accompanying drawings.

While the invention will be described in conjunction with the preferred

- 5 embodiments, it will be understood that they are not intended to limit the invention to these embodiments. On the contrary, the invention is intended to cover alternatives, modifications and equivalents, which may be included within the spirit and scope of the invention as defined by the appended claims. Furthermore, in the following detailed description of the present
- 10 invention, numerous specific details are set forth in order to provide a thorough understanding of the present invention. However, it will be obvious to one of ordinary skill in the art that the present invention may be practiced without these specific details. In other instances, well known methods, procedures, components, and circuits have not been described in detail as not to obscure aspects of the present invention unnecessarily.
- 15

Embodiments of the present invention are directed towards to an integrated dynamic dispersion compensating Fourier filter capable of balancing the dispersion introduced through various building blocks of a

20 communications network. The present invention provides dynamic dispersion compensation capable of handling not only static network conditions, but also changing network conditions. The present invention is more easily manufactured than prior art solutions and does not require very tight critical dimension control. The present invention and its benefits are

25 further described below.

Figure 5 shows a fiber-optic communication system in accordance with one embodiment of the present invention. As depicted in Figure 5, a transmitter 501 is used to transmit optical signals across a distance to a filter 502 and a receiver 503. In this embodiment, the filter 501 and the receiver 5 503 are integrated into a single PLC device 510. The receiver 503 can be any of a number of PLC devices designed to receive the transmitted optical signals and process them. Examples include arrayed waveguide grating devices (AWGs), optical signal detectors, and the like.

10 The filter 501 functions by creating a phase compensation profile and adding the phase compensation profile to the incoming signal from transmitter 501. As is well-known, there exists chromatic dispersion problems in the transmission of optical signals across long distances through fiber-optic cables. Chromatic dispersion is an important issue in high-speed 15 fiber optic communication networks. Such high-speed communication networks are often relied upon for transmitting terabytes of data across great distances. Data transfer capacity is in constant demand, and as new capacity is added to the network, that capacity is quickly used. One method of increasing data transfer capacity involves the increasing of the bit rate of the 20 transmission. As the bit rate of a transmission system increases, the susceptibility to chromatic dispersion increases. For example, in a fiber-optic transmission system functioning at 40Gb per second per channel, the system can only tolerate a dispersion on the order of 10 ps/nm.

25 Filter 501 functions as a Fourier filter that adds an excess phase (e.g., a phase compensation profile) to the incoming signal to compensate for the phase error introduced by dispersion. For example, an ideal dispersion

compensator requires change of phase without affecting the magnitude of the signal (e.g., no excess loss added to the signal). This constitutes a transfer function $h(f)$ that has the amplitude of unity at the absolute value of $h(f)$ and the desired phase change required is expressed as $\phi = \text{angle}(h(f))$. The
5 Fourier transform of this function is the required transfer function in z domain $H(z)$ that can be expanded in terms of its unit delays with complex coefficients, shown as equation 500 in Figure 5.

Figure 6 shows ϕ in diagram of two signal pulses received at the filter
10 501 and the two signal pulses after filtering by filter 501. The upper trace shows the two pulses having been chromatically dispersed such that their energy (on the vertical axis) overlaps one another with respect to time (on the horizontal axis). Such pulses can be difficult or impossible to reliably detect (e.g., using an optical detector). The lower trace shows the two pulses after
15 filtering and dispersion compensating within filter 501. As described above, filter 501 compensates for the chromatic dispersion, thereby providing more distinct separation between the two pulses, as shown.

In accordance with the present invention, the filter 501 functions by
20 ensuring chromatic dispersion remains within specified limits, even as bit rates of the transmission system increase or as transmission conditions within the network change (e.g., as new nodes are added or removed, new lengths of fiber-optic cable are added, etc.). Hence, since the transmission and dispersion characteristics of the network change as the network
25 reconfigures, filter 501 performs the important function of providing a compensation means not only for static transmission and dispersion conditions but also dynamic transmission and dispersion conditions.

Figure 7 shows a tap delay filter 700 in accordance with one embodiment of the present invention. Tap delay filter 700 is used to implement the Fourier filter function of the present invention. As depicted in Figure 7, filter 700 incorporates 4 delay lines 701-704. Mach Zehnder couplers 711-722 are shown. Thermal optic phase shifters 721-745 are shown.

As described above, the Fourier transform of the phase compensation transfer function in z domain $H(z)$ that can be expanded in terms of its unit delays with complex coefficients, shown as equation 500 in Figure 5. The unit delays can be implemented using delay taps. Referring still to equation 500 of Figure 5, a_n is the coefficient corresponding to a given delay line. In this embodiment, this coefficient can have amplitude and phase, where the amplitude, for example, defines the amplitude redistribution between different delay lines and the phase, for example, defines the phase adjustment. This phase adjustment can be implemented using, for example, thermal optic phase shifters. In either case, the function of equation 500 can be realized using a tap delay filter, such as tap delay filter 700.

20

Referring to Figure 7, the resolution of the filter 700 in compensating phase is determined by the number of tap delay lines that are used in the implementation. The free spectral range of the filter 700 is defined by the order of the path difference used between adjacent arms 701- 704. The method used in filter 700 involves the use trees of Mach Zehnder couplers 711-716 to distribute power between different arms 701-704 and recombine

them. An array of phase shifters 736-739 is used to introduce required phase shift for each arm.

Accordingly, the object of filter 700 is to allow incoming light to arms

5 701-704 with one wavelength to pass through unchanged, or with little attenuation, while light at other wavelengths is attenuated. Filter 700 shows an implementation of a tap delay filter. In tap delay filters, also referred to as Fourier filters, the filter function is achieved by breaking the light into beams that propagates different distances (e.g., through the adjacent arms
10 701-704). The light through each of these beams experience different phase change depending on the length of the arm and the particular wavelength. Hence, for a particular wavelength, a case can be realized wherein that light from all the arms interfere constructively and will pass through the filter. However, there exist other wavelengths wherein light from different arms
15 will have opposite phases that result in destructive interference. The extreme case of attenuation occurs when light from different arms are completely out of phase with each other cause a complete attenuation.

Thus, at the inputs of arms 701-704, light has the same phase. By the
20 end of the arms 701-704, the relationship (e.g., wavelength λ) between the light in the arms is such that light having a constructive phase relationship interfere constructively, while the light having a destructive phase relationship interferes destructively. By adjusting ΔL (e.g., by using phase shifters 736-739), the wavelengths that interfere constructively or
25 destructively can be selected in order to tune the transmission spectrum of the filter.

Referring still to figure 7, in a manner similar to tuning the power spectrum, filter 700 can effect the phase relation between different wavelengths propagating through arms 701-704 as well. As described above, although the wavelengths in arms 701-704 start out with, for example, a $\pi/4$ phase difference, at the output of the filter, the wavelengths show a phase difference of 0. The output phase corresponds to the phase of the net beam emerging from different arms. If the incoming beam carries pulsed information, the resulting difference in phase will translate into different traveling time for pulses with different wavelengths. This traveling time difference is known as group delay. Chromatic dispersion is defined as the variation of group delay between different wavelengths.

The Fourier filter 700 compensates for chromatic dispersion by generating an "inverse" chromatic dispersion. When the light propagates through the filter 700 and is separated between different arms 701-704, the phase relation between light with different wavelengths changes. The emerging light depends on the net of all the lights from different arms 701-704 and how they add together. Therefore, the filter 700 can introduce a transmission spectral and also dispersion to the incoming beams. Usually, in Fourier filters the transmission spectral and phase are inter-related, as one can imagine by controlling the amplitude of the light that goes through a particular delay line, one could subtract (add) more to the transmitted amplitude as well as phase.

The phase response of the Fourier filter 700 is exploited to dynamically compensate for dispersion resulted from other sources in an optical network. Since a Fourier filter is constructed from a number of arms with different

delays (delays are geometrical series of a constant delay), the response of the filter can be shown mathematically as (for ease of presentation we showed the transfer function in Z-domain):

$$H(z) = \sum a_n z^{-n}$$

- 5 where n defines arm with a delay n times the initial delay, and a_n is the complex amplitude of light that goes through that arm. Based on what we presented before, this can be separated into the amplitude transmission and phase as:

$$H = |H| e^{j\phi}$$

- 10 where $|H|$ is the amplitude response of the filter and ϕ is the phase response of the filter. It is desirable to have control over ϕ with minimum effect on $|H|$. In general, this is related to free spectral range (FSR) of the filter (e.g., the constant delay that delay arms are multiple of that). As the FSR is reduced (e.g., increase the delay between the arms), a smaller change yields a large change in phase, and consequently on group delay and dispersion. For example, filter 700 can generate a variable ± 20 ps/nm dispersion and change the slope of an input dispersion completely. One application would be the use of filter 700 to cancel the dispersion introduced by network conditions to a 100GHz DWDM channel.
- 15

20

Thus, embodiments of the present invention are directed towards to an integrated dynamic dispersion compensating Fourier filter capable of balancing the dispersion introduced through the various building blocks of a communications network. The present invention provides dynamic dispersion compensation capable of handling not only static network conditions, but also changing network conditions. The present invention is

more easily manufactured than prior art solutions and does not require very tight critical dimension control.

The foregoing descriptions of specific embodiments of the present
5 invention have been presented for purposes of illustration and description.
They are not intended to be exhaustive or to limit the invention to the precise
forms disclosed, and obviously many modifications and variations are
possible in light of the above teaching. The embodiments were chosen and
described in order best to explain the principles of the invention and its
10 practical application, thereby to enable others skilled in the art best to utilize
the invention and various embodiments with various modifications as are
suited to the particular use contemplated. It is intended that the scope of the
invention be defined by the claims appended hereto and their equivalents.

15